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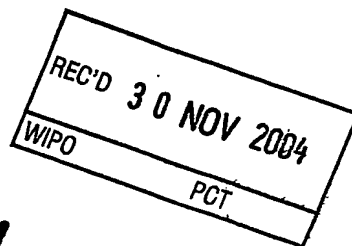
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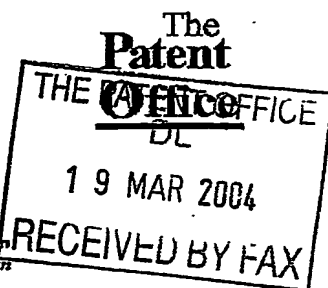
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1/77

**The Patent Office**Cardiff Road  
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19 MAR 2004**2. Patent application number**  
(The Patent Office will fill in this p

0406219.6

19 MAR 2004

**3. Full name, address and postcode of the or of each applicant** (underline all surnames)Roke Manor Research Limited  
Roke Manor, Old Salisbury  
Romsey, Hampshire SO51 0ZN

Patents ADP number (if you know it)

05615455007

If the applicant is a corporate body, give the country/state of its incorporation

UNITED KINGDOM

**4. Title of the invention**

A method of controlling interference from a transmitter in one communication system to a receiver in another communication system

**5. Name of your agent (if you have one)**

Jan Payne

"Address for service" in the United Kingdom to which all correspondence should be sent (including the postcode)

Siemens plc  
Intellectual Property Department  
The Lodge, Roke Manor  
Romsey, Hampshire SO51 0ZN

Patents ADP number (if you know it)

4095873005

**6. Priority.** Complete this section if you are declaring priority from one or more earlier patent applications, filed in the last 12 months.Country Priority application number  
(if you know it)  
GB 0323429.1Date of filing  
(day / month / year)  
07.10.2003**7. Divisionals, etc:** Complete this section only if this application is a divisional application or resulted from an entitlement dispute (see note f).**8. Is a Patents Form 1/77 (Statement of inventorship and of right to grant of a patent) required in support of this request?**

Yes

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9. Accompanying documents: A patent application must include a description of the invention. Not counting duplicates, please enter the number of pages of each item accompanying this form.

Continuation sheets of this form

Description	14
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Abstract	1
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10. If you are also filing any of the following, state how many against each item.

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11.

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Date

Jan Payne  
Chartered Patent Agent

19.03.2004

12. Name and daytime telephone number of  
person concerned in the United Kingdom

in Patent

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DUPLICATE

1

A METHOD OF CONTROLLING INTERFERENCE FROM A TRANSMITTER IN  
ONE COMMUNICATION SYSTEM TO A RECEIVER IN ANOTHER  
COMMUNICATION SYSTEM

5 This invention relates to a method of controlling interference from a transmitter in one communication system to a receiver in another communication system. In general, a certain frequency band is allocated for a particular type of communication system, e.g. mobile phones and then within that frequency band, any equipment which wishes to communicate must share the allocated spectrum with other user equipment  
10 giving rise to the possibility of interference. In any system with at least one transmitter and one receiver, it is assumed that the transmitter is the cause of the interference. However, the fundamental problem of spectrum sharing is not the transmitters, but the receivers. This is because, with current systems, it is possible to sense the presence or absence of signal from a transmitter that is already operating on a frequency one wishes  
15 to share, but it is not possible to determine whether transmission on a particular frequency will cause interference at a receiver or not. The presence of a strong signal might indicate that a remote receiver will have no problem with any interference the transmitting terminal might generate. Equally a weak signal might indicate that a nearby receiver cannot cope with any interference the transmitting terminal might  
20 produce. This runs contrary to the standard philosophy of dynamic channel assignment in which the channels with minimum interference are preferred.

In accordance with a first aspect of the present invention, a method of controlling interference from a transmitter in one communication system to a receiver in another communication system comprises transmitting a beacon from the receiver  
25 representative of a frequency at which the receiver is trying to receive; listening for the beacon at a beacon receiver associated with the transmitter; and deriving a power spectral density limit for a transmission from the transmitter based upon the strength of the beacon received at the beacon receiver.

The problem of whether or not a transmission will cause interference can only  
30 be addressed if it is known where the receivers are, so in the present invention, receivers transmit and transmitters receive to convey this information.

Preferably, for a plurality of beacons received representing the same frequency, the derived maximum transmit power spectral density is related to that of the beacon received at the highest power.

5 It is assumed that if a transmission is set, such that it does not interfere with the receiver represented by the beacon at the highest power, then all the others will be unaffected too, whether because they are further away or otherwise more tolerant of interference.

10 Preferably, the method further comprises comparing the derived transmit power spectral density limit with a predetermined minimum transmit power spectral density required by the transmitter for that frequency; and transmitting a signal at that frequency, only if the determined transmit power spectral density limit exceeds the minimum.

15 If the transmission would not be of sufficient power to reach its intended destination, then it is not sent. This can be overcome by sending via an intermediate terminal, so that the required minimum level is lower.

Preferably, a predetermined maximum transmit power spectral density is set, if no beacons are received at the transmitter.

20 Preferably, the method further comprises choosing a transmission frequency for the transmitter which permits the maximum power spectral density for the transmission.

25 Preferably, the transmission from the transmitter is transmitted at a frequency derived by determining the strongest received beacon which represents any one frequency; thereafter selecting, from the determined strongest beacons, the beacon with the lowest power; and transmitting at the frequency represented by that selected beacon.

Preferably, a transmit power spectral density for a transmission from the transmitter is set dependent upon the strength of the received beacon at the chosen frequency.

Preferably, the maximum permitted power spectral density of the transmitter is

minimum signal to noise ratio for receiving the beacon in its effective bandwidth and the noise figure of the beacon receiver at the transmitter.

Preferably, a random time division multiple access (TDMA) protocol is applied, whereby beacons representing different frequencies transmit at different times, such that over a series of cycles a beacon representing each frequency will be heard at a different time relative to another particular represented frequency, such that no one frequency at a higher power consistently blocks reception of a beacon representing another frequency at a lower power.

Preferably, a code division multiple access (CDMA) protocol is applied, whereby beacons representing different frequencies are distinguished from one another by different codes.

Preferably, a correlation period of a CDMA component of the beacon signal is controlled by an FFT controller.

Preferably, each beacon transmits a type identifier and each beacon receiver comprises type specific correlation means, such that a beacon receiver can ignore same type beacons in determining whether or not or how much power to transmit.

Preferably, a receiver transmits a beacon only if interference levels exceed an acceptable value.

Preferably, the beacon power is adapted to the wanted signal power received at the receiver.

Preferably, the beacon power is adapted to the interference power received at the receiver.

Preferably, a bandwidth managed by a beacon is sufficiently narrow that substantial correlation of shadow fading applies across that bandwidth.

Preferably, beacon communication is separated from a spectrum that is managed by the beacons by using a different frequency.

Preferably, each beacon occupies a frequency bandwidth which is small compared with the total bandwidth managed by that beacon.

Preferably, neighbouring beacons in a managed bandwidth manage discrete contiguous sections of frequency, each section comprising a fraction of the beacon managed band, each beacon being separated from the frequency bandwidth which it manages by the alternate fraction.

The fractions can be unequal, such as  $1/3$  and  $2/3$ , but typically each fraction is  $1/2$ .

Preferably, the beacon receiver is periodically tested with an internal beacon of known power and its associated transmitter is prevented from transmitting if a beacon receiver fault occurs.

Alternatively, beacon and system communications are separated in time.

This requires timing control, for example, provided by putting satellite timing receivers in all transmit and receive terminals.

Preferably, beacon reception and transmission happen at the same equipment, separated in time, by arranging for reception to take place whenever transmission is not according to schedules of the random TDMA protocol.

In accordance with a second aspect of the present invention, a communication system comprises at least one transmitter and at least one receiver, wherein a beacon transmitter is associated with the at least one receiver and a beacon receiver is associated with the at least one transmitter, whereby a power spectral density for transmission at any one transmitter is determined based upon the strength of the or each beacon received at the associated beacon receiver.

In accordance with a third aspect of the present invention, a transmitter for a communication system, the transmitter being provided with an associated beacon receiver, whereby a power spectral density for transmission from the transmitter is determined based on the strength of one or more beacons received at the associated beacon receiver.

In accordance with a fourth aspect of the present invention, a receiver for a communication system, the receiver being provided with an associated beacon transmitter, whereby a beacon can be transmitted by the beacon transmitter to control interference, such that a desired maximum power spectral density of interference received at the receiver is achieved.

An example of a method of controlling interference from a transmitter to a receiver in a communication system will now be described with reference to the

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Figure 2 illustrates a frequency band structure for the beacons and managed bands using the method of the present invention.

Fig. 1 illustrates one example where the method of the present invention may be applied. A first receiver 1 which is trying to receive from a first transmitter (not shown) needs to prevent interference 2 on the frequency that it is trying to receive. To do this it transmits a beacon 3 representing that frequency. This beacon is received at a beacon receiver associated with a second transmitter 4 which is trying to transmit to a second receiver (not shown). If the second transmitter 4 transmits it will produce interference 2 at the first receiver 1.

The beacon 2 transmitted from the first receiver represents the frequency at which the receiver 1 is trying to receive. Clearly this cannot be the same frequency as that being received, so a nearby frequency is assigned for this purpose. A compromise is required, so that the frequency used is near enough to have good correlation of path loss, but far enough away to allow duplexing.

Although Fig.1 only shows one receiver and one transmitter, in a practical system there are multiple receivers and transmitters. The basic principle is that some or all of the receivers 1 transmit a beacon 3 at a suitable power level. Any terminal with a transmitter 4 that can hear a beacon related to a given frequency must reduce its transmitter power according to the strength of the beacon. If the permissible power is too low to allow the required communications then the terminal cannot use that frequency. A terminal operating dynamic channel allocation (DCA) would scan around the beacons and select the frequency whose beacon was received at the lowest power. If multiple beacons representing the same frequency are received at different powers the power of interest is taken to be the strongest.

The beacon transmission band is arranged as a band of frequencies at either end (or both ends) of the band being managed. There is one separate beacon signal relating to each frequency in the band and these must be multiplexed together. Determination of beacon power and interference limitation is explained below for the situation illustrated in Fig.1. In the two links, primary status has been allocated to a terminal trying to receive and secondary status to the one trying to transmit and for simplicity the transmitter for the primary link and the receiver for the secondary link are not



shown. The primary receiver 1 is transmitting its beacon 3 in order to guarantee reception.

The following assumptions are made:

- $P_P$  is the primary receiver beacon power
- 5  $P_S$  is the secondary transmitter power spectral density
- $g$  is the path gain between the two equipment connectors – this includes propagation path gain ( $\ll 1$ ), antenna gains and feeder losses
- $N_P$  is the primary receiver noise figure
- $N_S$  is the noise figure of the beacon receiver at the secondary transmitter
- 10  $B_S$  is the effective bandwidth at the beacon receiver for receiving the beacon
- $\gamma_S$  is the minimum signal to noise ratio for receiving the beacon in its effective bandwidth
- $d$  is the factor by which we allow the primary receiver to be de-sensitised

- The level of  $P_S$  needs to be set such that it is measurable at a point where a
- 15 secondary transmitter could produce non-trivial interference at the primary receiver. This is done by fixing the maximum power that a secondary transmitter can emit when the beacon is undetectable, such that only acceptable interference arises.

The limit sensitivity for the beacon receiver is  $L_S = kTB_S N_S \gamma_S$

- (where  $k$  is Boltzmann's constant and  $T$  is the operating temperature in degrees
- 20 Kelvin).

- Thus, if the received beacon power is less than or equal to  $L_S$ , then the secondary transmitter will emit a power spectral density of  $P_{S-MAX}$  Watts/Hz. This is set so that if the beacon signal is equal to  $L_S$ , then only acceptable interference will be generated at the primary receiver. Assuming that this condition happens for a path gain
- 25 of  $g$ , then the received beacon power will be given by

$$L_S = P_P \cdot g. \text{ Thus } P_P \cdot g = kTB_S N_S \gamma_S.$$

The interference power spectral density generated at the primary receiver will then be  $P_{S-MAX} \cdot g$ . The acceptable level for this will be  $kTN_P(d-1)$ . Thus  $P_{S-MAX} \cdot g = kTN_P(d-1)$ .  
Eliminating  $g$  from the two equations gives

In practice, for universal operation, a reference needs to be set. For example, if the reference is set at 1W for a de-sensitisation of 3 dB with a 0 dB noise figure primary receiver, then conveniently this gives a beacon power for 3 dB de-sensitisation equal to  $1/N_P$  Watts.

- 5 In the following example, the permitted de-sensitisation is 3 dB, both receivers have a noise figure of 6 dB,  $B_S = 1$  kHz and  $\gamma_I = 10$  dB. Thus, the beacon power is  $\frac{1}{4}$  Watt and so  $P_{S-MAX} = 2.5 \times 10^{-5}$  W/Hz.

- 10 In this example, a secondary transmitter with a bandwidth, for example, of 100 KHz is allowed to transmit up to 2.5 W if it cannot hear a beacon. As the beacon is detected and its received power climbs above  $L_S$ , then the maximum allowable power has to be reduced pro-rata.

- 15 From this simple analysis it can be seen that the appropriate beacon power is independent of primary system bandwidth, operating range, range to the secondary transmitter(s) and the primary system required signal to noise ratio. The beacon power depends only on the primary receiver noise figure and the allowable de-sensitisation. Once a reference for beacon power has been legislated, only relatively small variations will result. The reference power, assuming receiver noise temperatures no lower than  $290^\circ$  K and desensitisation by no less than 3 dB, will be the maximum beacon power ever used.

- 20 Feeder loss and antenna patterns do not affect the beacon power or maximum secondary power conditions for the beacon system, although they do affect the spectrum sharing that is available. The maximum power a secondary transmitter may transmit depends on the sensitivity of its beacon receiver. Thus the equipment manufacturer is motivated to produce beacon receivers with low noise figure and good  
25 beacon detectability.

- Fig. 2 illustrates an arrangement of a beacon band structure for the method of the present invention. A beacon signal represents the reception of a particular frequency in a band. At least one beacon type is required for every frequency. It is necessary to determine how large a beacon managed band 8, 9 ought to be, how much spectrum  
30 should be allocated to beacon transmission and how many frequencies should be allocated within a beacon managed band. As in all tradeoffs the result is a compromise. On the one hand, it is desirable that the beacon managed band is as large as possible to reduce the relative overhead of the beacon channel, but on the other hand it should be

as narrow as possible in order to guarantee the same radio propagation characteristics across the band and, in particular, between any sub-band and the beacon transmission band.

As it is not practical to reduce the managed bandwidth down to the multipath fading correlation bandwidth of the channel, then the beacon transmission band must be wide enough to have reasonable inherent multipath diversity. If this beacon reception diversity is greater than that available to the users of the band, then the interference sensing mechanism will be conservative, i.e. it will tend to overestimate the interference generated. In the alternative situation, an underestimate may occur, so it may be necessary to introduce a fading margin to account for this difference. This can be done by transmitting the beacon at slightly higher power.

Based on the above requirement, it is desirable that the beacon managed band is no greater than about 5% of its centre frequency, which gives reasonable correlation of average propagation characteristics. Some care is needed to ensure that no tight frequency selectivities in the antennas cause difficulties. Such a problem is more likely to arise in the primary service, which will probably be incumbent, than in the secondary service, since the secondary service is generally designed to use any of the frequencies in the beacon managed band. One solution to the problem for the primary service is to bias the beacon transmitted power according to the relative antenna gains in the beacon transmission band and the represented signal frequency. Another feature is that the beacon transmission band is preferably set so that it does not exceed about 5% of the beacon managed band in order to avoid there being an unacceptable overhead for the beacons.

Consideration must be given to the provision of duplexing filters since the primary receiver must transmit and the secondary transmitter must receive. Clearly if the beacon must be transmitted to represent a frequency that is right next to the beacon transmission band there will be a problem. This difficulty can be solved by having beacon transmission bands 5, 6, 7 at each end of the beacon managed band 8, 9. The frequencies available for use are then represented by the beacon transmission band

band 6 also manages frequency range 12 of beacon managed band 9 whilst band 7 manages frequency range 13. The size of the first and second range 10, 11 is generally equal, i.e. half of each managed band is managed by one beacon transmission band 5 and half by the other 6, but these ranges can be different proportions of the total beacon managed band 8, such as 1/3 and 2/3 or other ratios.

In a specific example, a frequency band is set around 2 GHz, with a bandwidth 8 of 100 MHz. This leads to a beacon transmission band 6 of 5 MHz (based on 5% bandwidth). According to the structure of Fig. 2, there is always a frequency spacing of at least 50 MHz to allow for duplexer filtering.

The choice of number of frequencies represented is again, a trade-off between complexity and flexibility. From the viewpoint of simplicity we would choose to have as few as possible. For example, if our 100 MHz wide band had an incumbent system with 5 MHz channels then the obvious answer would appear to be to have twenty, 5 MHz channels. However, this pre-supposes that there will be only one further tier of spectrum sharing and that the sharing system will also have a bandwidth of 5 MHz or multiples thereof. It may be that some sharing systems are introduced based on 7 MHz. In this case they may need to transmit as many as 3 beacons to keep a channel free. This would reserve 15 MHz of spectrum which would be inefficient. From the viewpoint of flexibility the answer would be to use the highest common divisor of all anticipated channel spacings. The disadvantage of this approach is that it might lead to the need for a large number of beacons for each frequency used. For example, a spacing of 1 MHz would require 5 beacons to keep a 5 MHz channel free. The number of frequencies would thus be a compromise between wasted spectrum and having a large number of beacons. At this stage a bandwidth of 5 MHz nevertheless seems reasonable leading to 20 beacons in this example.

At any given beacon receiver there may be beacons from a number of receivers, relating to some different and some common frequencies. A would-be transmitter wants to examine beacons for a particular frequency. If there are many beacons for that frequency it is desirable that the task should not grow unduly. The options for multiple access are division in frequency, time and code (FDMA, TDMA and CDMA respectively).

If beacon transmitters and receivers are to be low cost, then implicitly they have relatively inaccurate frequency references. For example, a 20 ppm end to end error

would not be unreasonable, but at 2 GHz this would correspond to an error of 40 kHz making any FDMA system difficult to operate. Additionally, each carrier would be relatively narrowband for FDMA. For example, representing 20 frequencies in 5 MHz gives a maximum channel spacing of 250 kHz and so the multipath diversity inherent in a bandwidth of 5 MHz is not exploited.

It is desirable for there to be no need for co-ordination between the different beacon transmitters, so no framing structures can be imposed, as required in TDMA. GPS receivers could be used to achieve this, but these can also have problems, particularly given the need for indoor operation and without some sort of timing control, the only available form of multiplexing in the time domain is random transmission, akin to the Aloha protocol. However, a problem with this is that *a priori* it is not possible to be sure how many beacons need to be received in a given area, so it is difficult to dimension the average duty cycle of the transmission.

CDMA allows for a very large number of beacons relating to the same frequency since they would have the same code and would be separated only in time, but this suffers from the near far problem. If a strong beacon relating to one frequency is received then the receiver may know that that frequency is unavailable. However, it cannot know whether other frequencies are available since the beacon receiver is desensitised and it must be assumed that other beacons *may* be present.

Overall, a combination of random TDMA and CDMA is preferred. The CDMA element provides the ability to measure multiple beacons relating to the same frequency and the TDMA element provides resistance to the near far effect. By listening for long enough, a beacon receiver can accumulate adequate confidence that, if there had been a beacon for a particular receiver at a lower level than a strong signal then it would have had an opportunity to receive it. Where a beacon transmitter needs to send multiple signals from one location (e.g. for a frequency that requires several contiguous sub bands) they must all be transmitted simultaneously and not in sequence. This is to avoid a single source occupying too much of the time.

It is desirable that the beacon receiver is made as sensitive as possible, which

over which it is usually possible to correlate is one half cycle of the error frequency, in this case 12.5  $\mu$ s. This leads to an effective bandwidth of 80 kHz which is too wide for sensitive reception. This can be radically improved by the use of an FFT correlator, for example as described in D J R Van Nee & A J R M Coenen, "New Fast GPS Code-  
5 Acquisition Technique Using FFT", IEE Electronic Letters, 17<sup>th</sup> January 1991 Vol 27, pp158 - 160. By choosing the number of FFT bins, a correlation period of any length can be made. The maximum useful correlation period depends on the correlation time of the channel. This varies from about 200  $\mu$ s to 1 ms depending on the mobility of the transmitter, receiver or both. The underlying code can be set to have a duration of  
10 around 200  $\mu$ s with up to, for example, 4 repetitions allowed which can be used in a static receiver to improve sensitivity. This gives an effective bandwidth from 5 kHz down to 1.25 kHz allowing good sensitivity. The preferred FFT size is 16, 32 or 64, all practical with today's technology for a 10 MHz sampling rate, although with changes in technology other sizes may be appropriate. There is some scalloping loss from the  
15 FFT, but this can be improved by randomising the frequency over  $\pm 0.5$  bins.

A further issue for the beacon information is that although it appears that it is only necessary to know the frequency represented by the beacon, it is also necessary to distinguish beacons transmitted from receivers in the same system as the transmitter, otherwise the transmitter never gets an opportunity to transmit. At any given time there  
20 are only a relatively small number of systems sharing any given band. Thus, only a small number of codes are needed to distinguish the different systems will be needed. Eight bits would normally be sufficient. Every beacon transmitter appends its system type identifier to each beacon and transmits this using the applicable spread spectrum code. This part is not related to the represented frequency, but is married to it by its fine  
25 frequency, i.e. the FFT bin selected and its timing. Terminals for a given system only need to correlate codes for their own system, so a single matched filter can be used here. Thus, there is no loss in sensitivity in determining whether a beacon came from a beacon receiver's own type of system. In this case, the beacon receiver ignores this particular signal when determining the power at which it could operate.

30 As stated earlier, for a given represented frequency the preferred power to use for the beacon is the largest of the received powers. However, in multipath, there may be several paths all relating to the same beacon transmission. Inevitably there will be some ambiguity in determining this situation, but this can be resolved in most cases by

only taking other multipath components from the same FFT bin output as that which identified the strongest output, which means that in the absence of significant Doppler the multipath components from any given beacon transmitter all fall on the same fine frequency. In addition, only multipath components are taken from a narrow window  
5 around the largest peak, which constrains the components taken to within the maximum delay span of the propagation medium. Finally, a maximum number of paths are taken to avoid including peaks of noise.

The description above sets out the situation for a primary and a secondary system in which the receivers of the primary system have beacon transmitters and the transmitters of the secondary system have beacon receivers, but this case gives no  
10 protection for the secondary system. A more general set of possibilities is shown in Table 1

Table 1

Type of System	Beacon Transmitter	Beacon Receiver
1. Primary (Dominant)	x	
2. co-operating	x	x
3. Tertiary (Unprotected)		x

15

In addition, beacon transmissions could be made responsive for systems that could tolerate brief loss of communications. In this case a receiver would only transmit a beacon when it encountered unacceptable interference. The interfering sources would then detect the beacon and either stop transmitting or reduce their power. It might in  
20 some cases be desirable for type 2 systems to receive and transmit essentially simultaneously on the beacon channel. This would apply if the dynamics of spectrum sharing were such that it was not appropriate for the terminal to stop transmitting the beacon when it was making its own system transmissions. The random TDMA transmission format for the beacon would allow this to happen. The choice of level of  
25 protection for a system would be a matter for a combination of circumstance and

take account of inaccuracies in RF gains, beacon transmitters etc. Regulation would seek to prevent manufacturers degrading their equipment performance requirements excessively at the expense of spectral efficiency and for this reason there may be a case for regulating an absolute maximum beacon transmitted power for a certain class of receiver. Further, it may be desirable, and in some cases, acceptable to increase the beacon transmitted power to take account of multipath fading that sometimes reduces the received beacon power without affecting the interference path

There are several fundamental approaches that can be applied to the setting of the beacon power. A fixed beacon power can be transmitted, based on the equipment design. This approach is robust, but not very flexible and does not always lead to the best protection for the receiving equipment against multiple interferers. Another option is to arrange for the primary receiver to estimate the received noise and set the beacon power accordingly. In this case, as more interferers share the frequency the interference rises. This rise reduces the margin for further increase in interference leading to an increase in the beacon power. This increase forces the interferers to reduce their power, or prevents additional interferers from operating, in order to maintain the receiver sensitivity. Alternatively, the primary receiver is arranged to set its beacon power according to the maximum acceptable level of interference given the signal that it is receiving. This approach clearly should not be used in conjunction with power control within the system, as the two systems will interact in complex and potentially undesirable or unpredictable ways. However, for fixed power systems this approach can be used to maximise spectrum sharing.

The fundamental beacon concept is flexible enough to allow any or all of the above approaches to apply for different systems operating in the same beacon managed band. In any given band, care needs to be taken in considering the types of system sharing the spectrum to ensure that overall stability is preserved. It would be possible, for example, to operate one of each of the types of system in Table 1 without instability. The type 1 system does not respond to beacons; the type 2 system defers to the type 1 system and protects itself from the type 3 system and the type 3 system defers to the types 1 and 2 systems and uses whatever spectrum it can find. For broadcast systems there are further degrees of freedom. For example, if one wished to use the beacon concept to allow sharing with the television band then every receiver



needs to be equipped with a beacon transmitter, however, a beacon only needs to be transmitted corresponding to the channel that the viewer is receiving at the time.

The strong desirability for transmitting the beacon through the same antenna as that used for reception, in order to share the antenna's properties, precludes the use of  
5 simple masthead pre-amplifiers. Instead, either a masthead pre-amplifier is arranged to be the source of the beacon transmission and the reference power chosen in such a way that the absolute power requirement is always modest, or else a diplexer is built into the masthead pre-amplifier with gain for the beacon transmitter as well.

Although these requirements add complexity, they do not create insuperable problems  
10 and additional revenue from the freed up spectrum should more than cover such costs.

The whole concept of spectrum sharing through beacons depends on the beacons being received whenever present with power above the presumed reference sensitivity. Equipment with a failed beacon receiver may transmit at maximum power on any frequency and cause arbitrary interference, so the status of the beacon receivers  
15 must be known at all times. To deal with this a built in self-test for the beacon receivers can be provided, in which a known beacon is injected periodically into the receiver at limit range level in order to test its function. Failure to receive the test beacon results in the equipment being prevented from transmitting.

CLAIMS

1. A method of controlling interference from a transmitter in one communication system to a receiver in another communication system, the method comprising transmitting a beacon from the receiver representative of a frequency at which the receiver is trying to receive; listening for the beacon at a beacon receiver associated with the transmitter; and deriving a power spectral density limit for a transmission from the transmitter based upon the strength of the beacon received at the beacon receiver.
2. A method according to claim 1, wherein, for a plurality of beacons received representing the same frequency, the derived maximum transmit power spectral density is related to that of the beacon received at the highest power.
3. A method according to claim 1 or claim 2, further comprising comparing the derived transmit power spectral density limit with a predetermined minimum transmit power spectral density required by the transmitter for that frequency; and transmitting a signal at that frequency, only if the determined transmit power spectral density limit exceeds the minimum.
4. A method according to any preceding claim, wherein a predetermined maximum transmit power spectral density is set, if no beacons are received at the transmitter.
5. A method according to any preceding claim, the method further comprising choosing a transmission frequency for the transmitter which permits the maximum power spectral density for the transmission.
6. A method according to any of claims 1 to 4, wherein the transmission from the transmitter is transmitted at a frequency derived by determining the strongest received beacon which represents any one frequency; thereafter selecting, from the determined strongest beacons, the beacon with the lowest power, and transmitting at the frequency represented by that selected beacon.

7. A method according to claim 5 or claim 6, wherein a transmit power spectral density for a transmission from the transmitter is set dependent upon the strength of the received beacon at the chosen frequency.

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8. A method according to any preceding claim, wherein the maximum permitted power spectral density of the transmitter is set at the product of the receiver beacon power; and a factor by which the receiver can be de-sensitised minus one; and the resultant of the receiver noise figure divided by the product of the effective bandwidth at the beacon receiver for receiving the beacon, the minimum signal to noise ratio for receiving the beacon in its effective bandwidth and the noise figure of the beacon receiver at the transmitter.

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9. A method according to any preceding claim, wherein a random time division multiple access (TDMA) protocol is applied, whereby beacons representing different frequencies transmit at different times, such that over a series of cycles a beacon representing each frequency will be heard at a different time relative to another particular represented frequency, such that no one frequency at a higher power consistently blocks reception of a beacon representing another frequency at a lower power.

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10. A method according to any preceding claim, wherein a code division multiple access (CDMA) protocol is applied, whereby beacons representing different frequencies are distinguished from one another by different codes.

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11. A method according to claim 10, wherein a correlation period of a CDMA component of the beacon signal is controlled by an FFT controller.

12. A method according to any preceding claim, wherein each beacon transmits a

any identification number of a beacon receiver transmits a beacon signal to a beacon receiver  
a beacon receiver transmits a beacon signal to a beacon receiver  
a beacon receiver transmits a beacon signal to a beacon receiver

13. A method according to any preceding claim, wherein a receiver transmits a beacon only if interference levels exceed an acceptable value.
14. A method according to any preceding claim, wherein the beacon power is adapted to the wanted signal power received at the receiver.
15. A method according to any of claims 1 to 13, wherein the beacon power is adapted to the interference power received at the receiver.
16. A method according to any preceding claim, wherein a bandwidth managed by a beacon is sufficiently narrow that substantial correlation of shadow fading applies across that bandwidth.
17. A method according to any preceding claim, wherein beacon communication is separated from a spectrum that is managed by the beacons by using a different frequency.
18. A method according to claim 17, wherein each beacon occupies a frequency bandwidth which is small compared with the total bandwidth managed by that beacon.
19. A method according to claim 18, wherein neighbouring beacons in a managed bandwidth manage discrete contiguous sections of frequency, each section comprising a fraction of the beacon managed band, each beacon being separated from the frequency bandwidth which it manages by the alternate fraction.
20. A method according to claim 19, wherein each fraction is  $\frac{1}{2}$ .
21. A method according to any preceding claim, wherein the beacon receiver is periodically tested with an internal beacon of known power and its associated transmitter is prevented from transmitting if a beacon receiver fault occurs.
22. A method according to any of claims 1-16 or 21, wherein beacon and system communications are separated in time.

23. A method according to at least claim 9, wherein beacon reception and transmission happen at the same equipment, separated in time, by arranging for reception to take place whenever transmission is not required according to schedules of the random TDMA protocol.

24. A method of controlling interference between a primary receiver and a secondary transmitter in a communication system as hereinbefore described with reference to the accompanying drawings.

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25. A communication network comprising at least one transmitter belonging to one communication system and at least one receiver belonging to another communication system, wherein a beacon transmitter is associated with the at least one receiver and a beacon receiver is associated with the at least one transmitter, whereby a power spectral density for transmission at any one transmitter is determined based upon the strength of the or each beacon received at the associated beacon receiver.

26. A transmitter for a communication system, the transmitter being provided with an associated beacon receiver, whereby a power spectral density for transmission from the transmitter is determined based on the strength of one or more beacons received at the associated beacon receiver.

27. A receiver for a communication system, the receiver being provided with an associated beacon transmitter, whereby a beacon can be transmitted by the beacon transmitter to control interference, such that a desired maximum power spectral density of interference received at the receiver is controlled.

## ABSTRACT

A method of controlling interference from a transmitter (4) in one communication system to a receiver (1) in another communication system, the method comprising transmitting a beacon (3) from the receiver representative of a frequency at which the receiver is trying to receive; listening for the beacon at a beacon receiver associated with the transmitter; and deriving a power spectral density limit for a transmission (2) from the transmitter based upon the strength of the beacon received at the beacon receiver.

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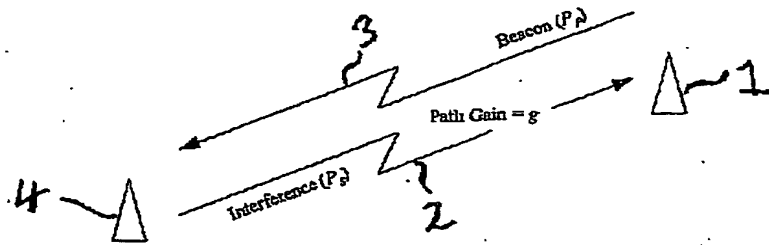


Figure 1 Beacon Transmission

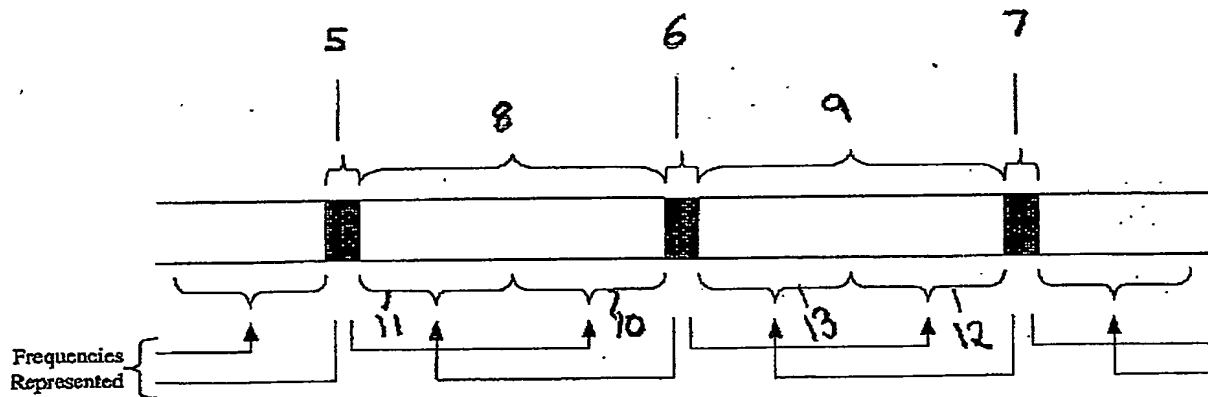


Figure 2 Beacon Transmission Structure

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